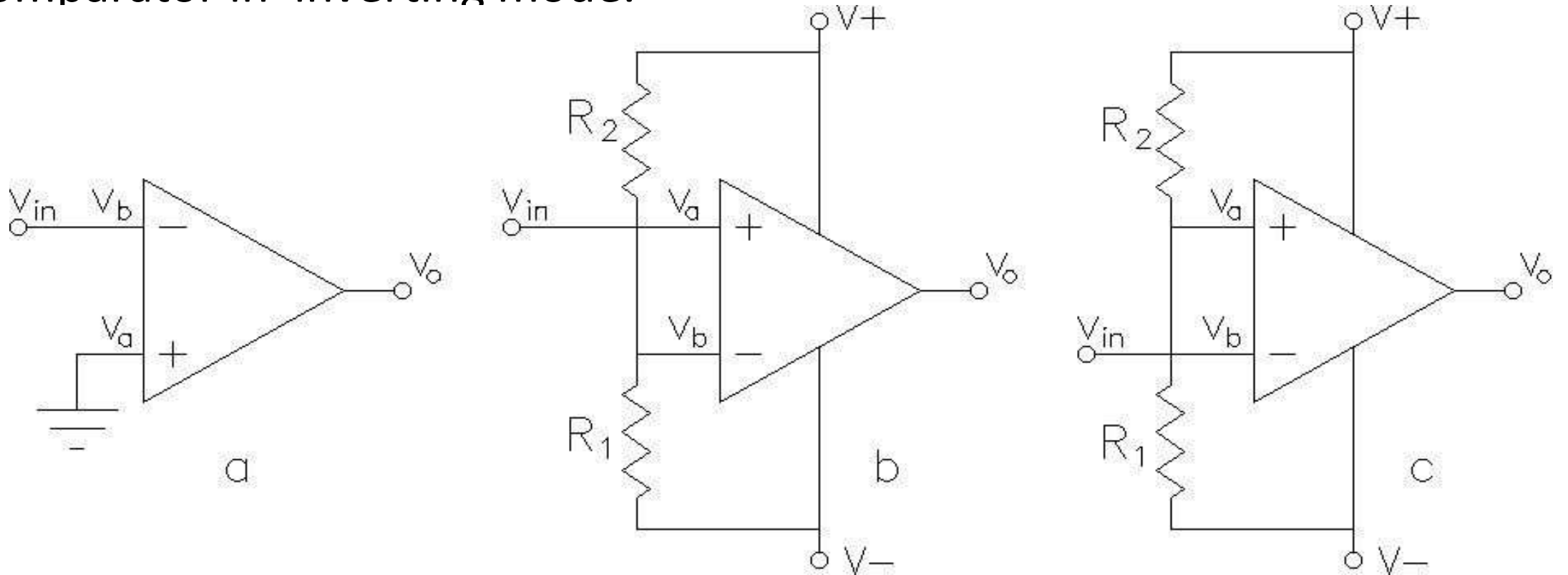


- **Module 3 : Non-Linear Applications of Op-amp**

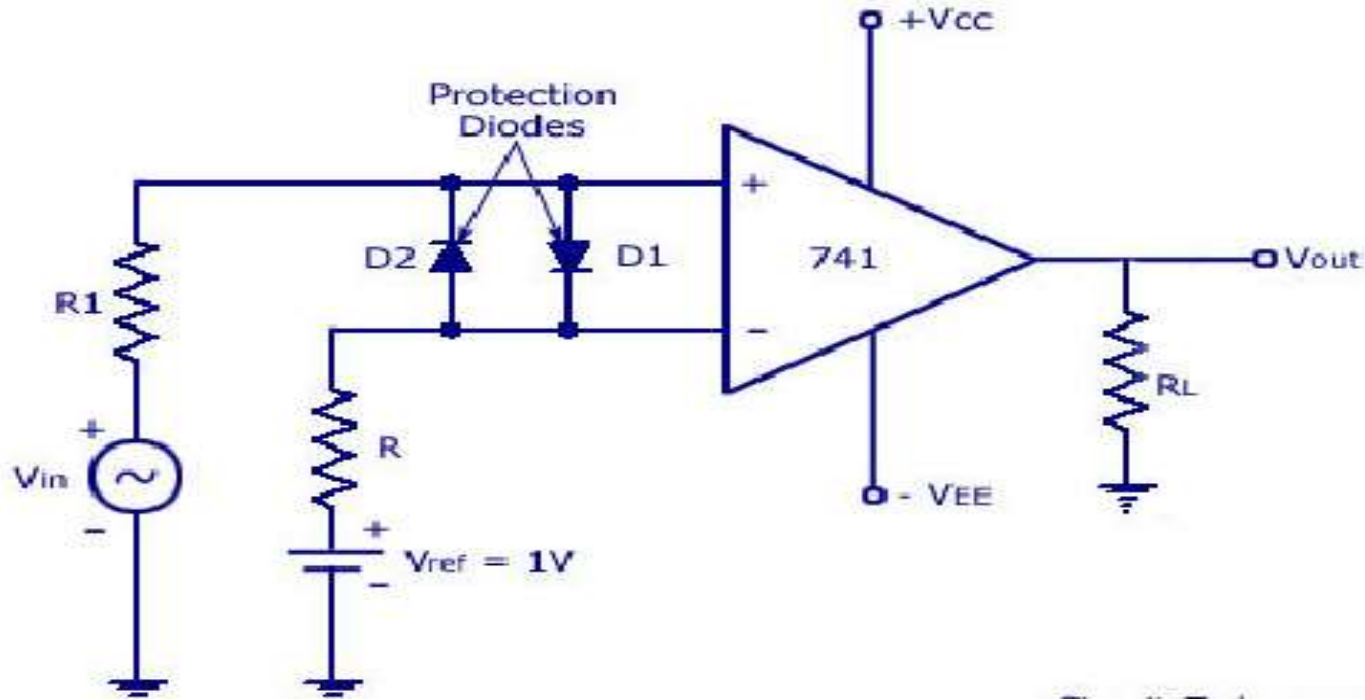
### 3.1 Comparators :

**A Voltage comparator circuit :** A Voltage comparator is a circuit which compares two voltages and switches the output to either high or low state depending upon which voltage is higher.

A voltage comparator based on op-amp is shown in fig a. Fig b shows a voltage comparator in non-inverting mode and Fig c. shows a voltage comparator in inverting mode.

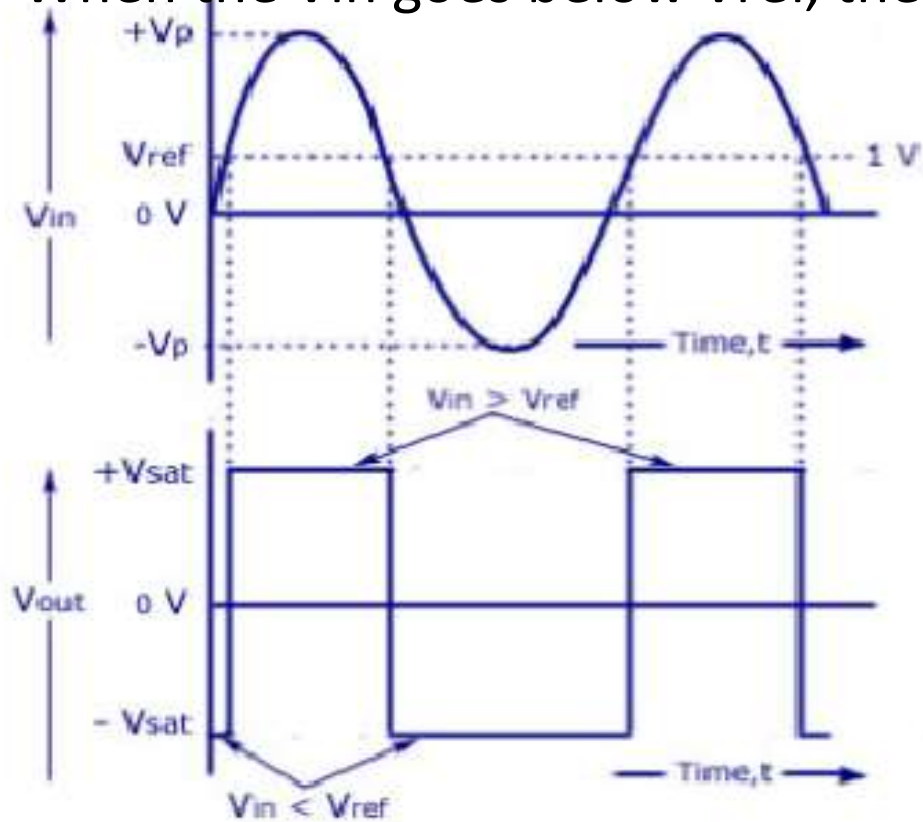


## Non-Inverting Comparator Circuit



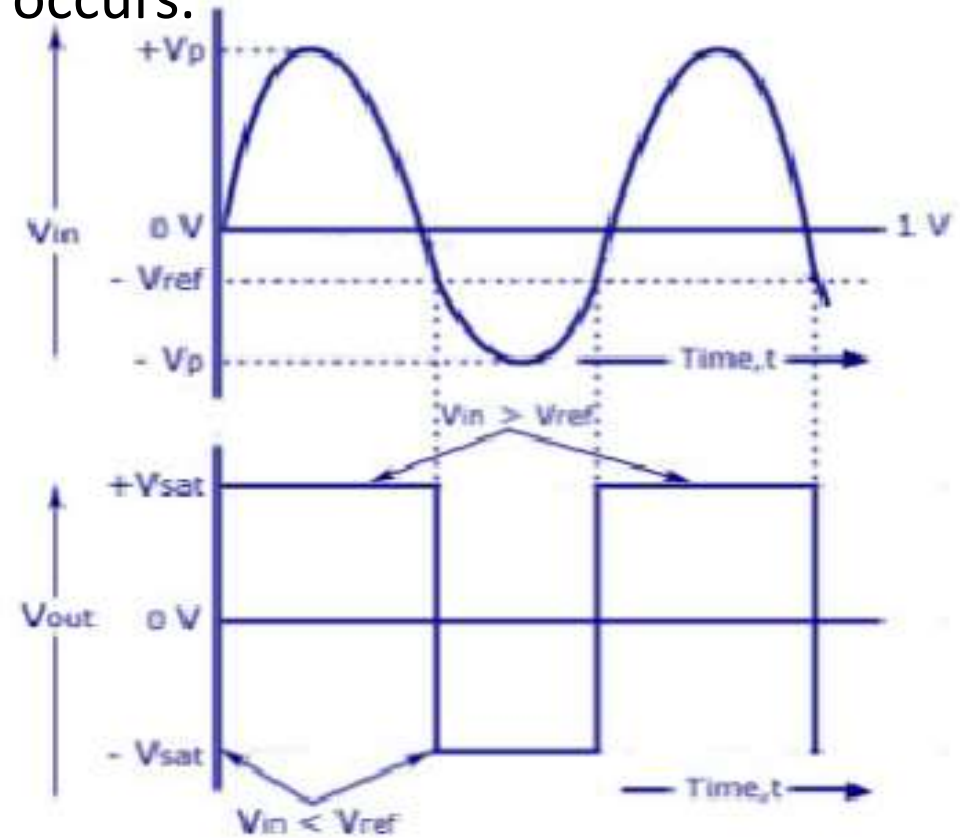
- In non-inverting comparator the reference voltage is applied to the inverting input and the voltage to be compared is applied to the non-inverting input. Whenever the voltage to be compared ( $V_{in}$ ) goes above the reference voltage, the output of the op-amp swings to positive saturation ( $V+$ ) and vice versa. Actually what happens is that, the difference between  $V_{in}$  and  $V_{ref}$ , ( $V_{in} - V_{ref}$ ) will be a positive value and is amplified to infinity by the op-amp.

- Since there is no feedback resistor  $R_f$ , the op-amp is in open loop mode and so the voltage gain ( $A_v$ ) will be close to infinity. So the output voltage swings to the maximum possible value ie;  $V_+$ .
- When the  $V_{in}$  goes below  $V_{ref}$ , the reverse occurs.



Input and Output Waveforms  
For Positive  $V_{ref}$

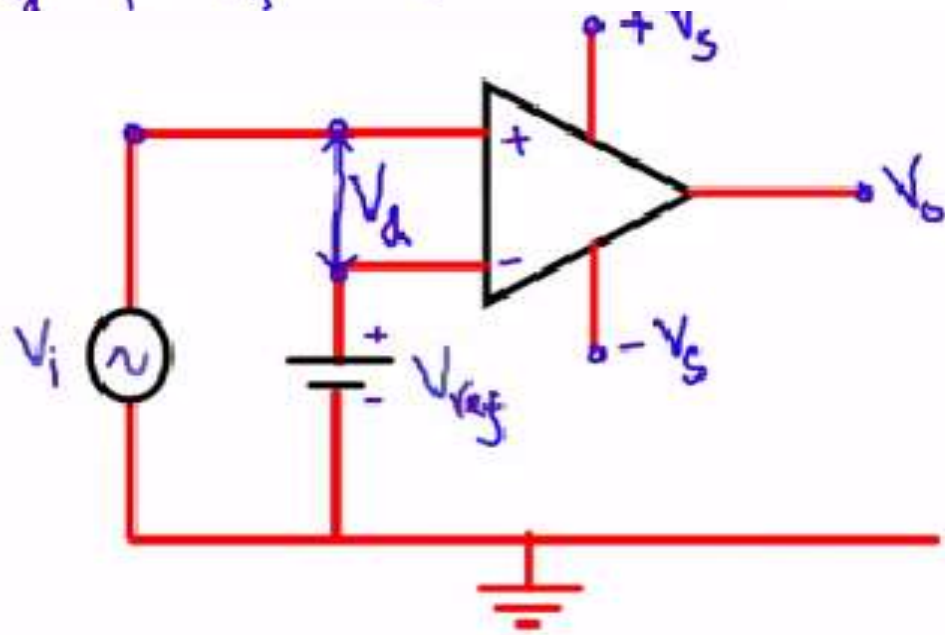
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Input and Output Waveforms  
For Negative  $V_{ref}$

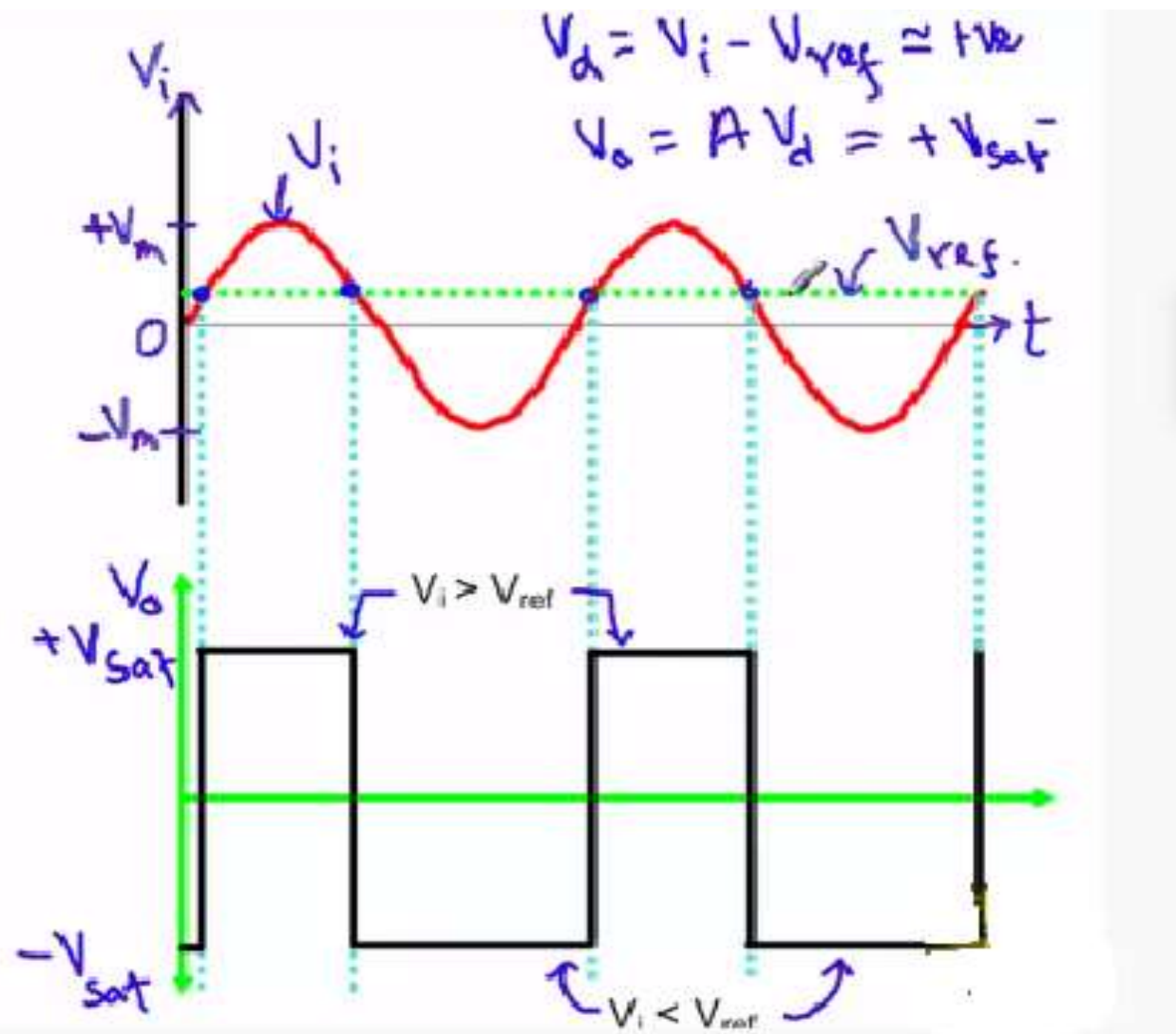
Op-Amp Non-Inverting Comparator:  $A \rightarrow 10^5$  to  $10^8$

$$V_d = V_i - V_{ref} \quad , \quad V_o = A V_d$$



$$V_i < V_{ref} \quad , \quad V_d = V_i - V_{ref} = -V_r$$

$$V_o = A V_d \approx -V_{sat}$$



Op-Amp as Non-Inverting Comparator

## Applications of a Comparator

- Zero crossing detector (ZCD), Window detector.
- Time marker generator, Phase detector.

❖ MCQ: If the input to the circuit of figure is a sine wave the output will be  
Options are: A) A Half-wave rectified sine wave    B) A Full-wave rectified sine wave  
C) Triangular wave    D) Square wave

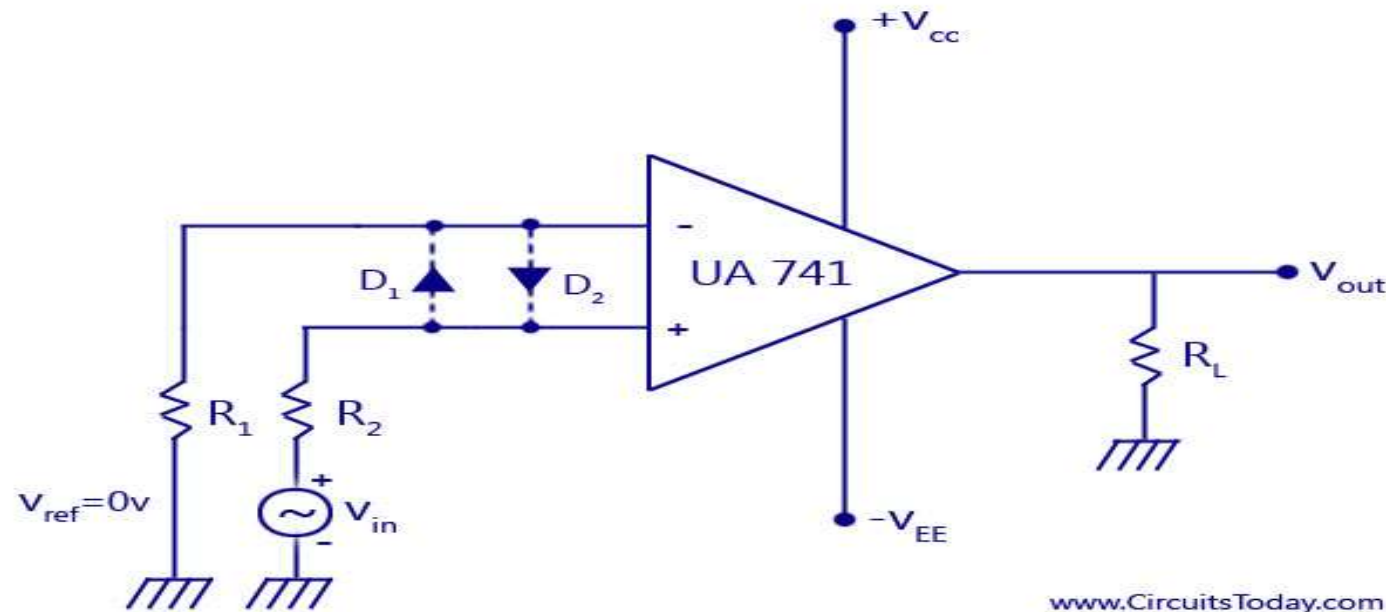
### Comparator Characteristics :

1. **Operation Speed** – According to change of conditions in the input, a comparator circuit switches at a good speed between the saturation levels and the response is instantaneous.
2. **Accuracy** – Accuracy of the comparator circuit causes the following characteristics:-
  - (a) **High Voltage Gain** – The comparator circuit is said to have a high voltage gain characteristic that results in the requirement of smaller hysteresis voltage. As a result the comparator output voltage switches between the upper and lower saturation levels.
  - b) **High Common Mode Rejection Ratio (CMRR)** – The common mode input voltage parameters such a noise is rejected with the help of a high CMRR.
  - (c) **Very Small Input Offset Current and Input Offset Voltage** – A negligible amount of Input Offset Current and Input Offset Voltage causes a lesser amount of offset problems.

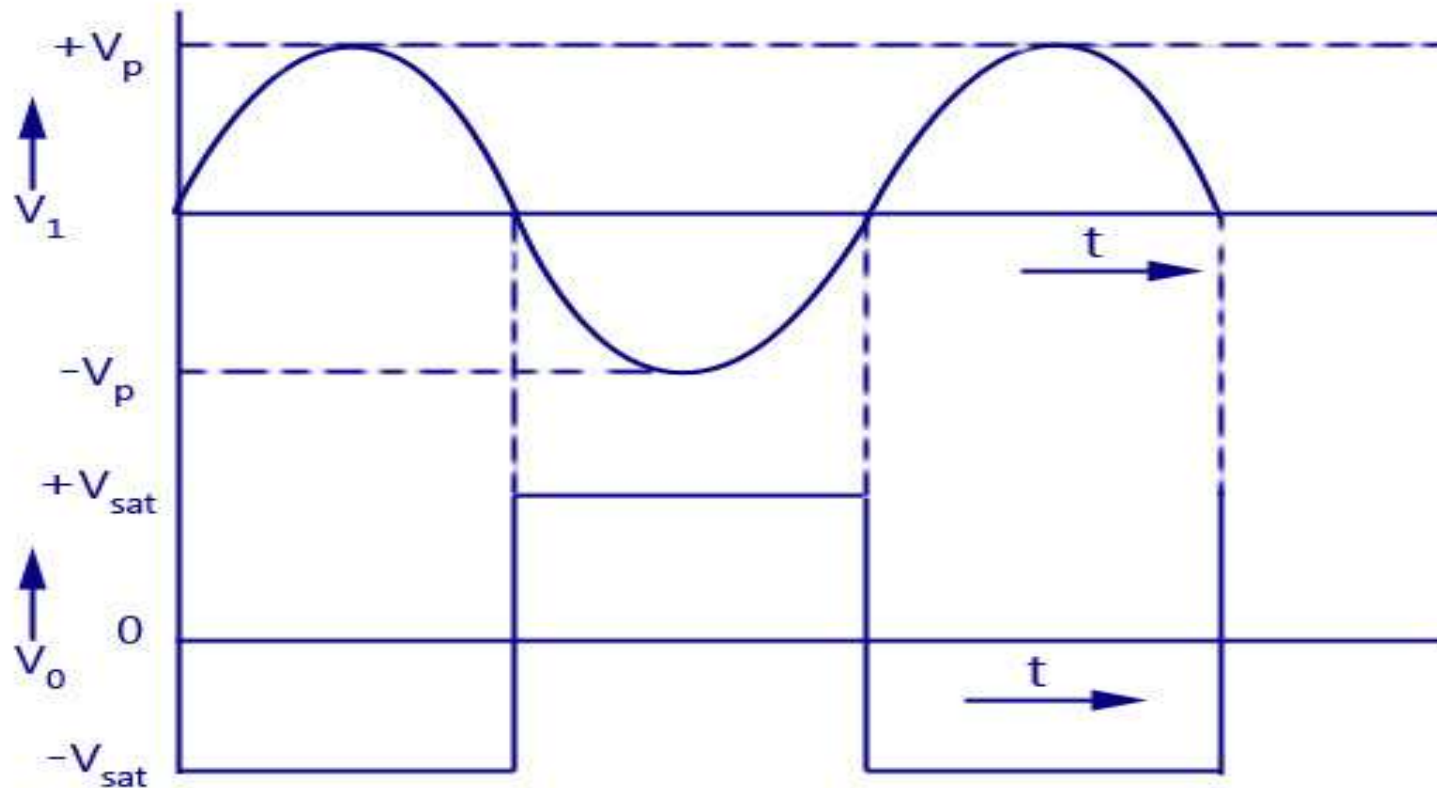
# 1. Zero crossing detector

- The zero crossing detector circuit is an important application of the [op-amp comparator circuit](#).
- It can also be called as the sine to square wave converter.
- Anyone of the inverting or non-inverting comparators can be used as a zero-crossing detector.
- The only change to be brought in is the reference voltage with which the input voltage is to be compared, must be made zero ( $V_{ref} = 0V$ ).
- An input sine wave is given as  $V_{in}$ . These are shown in the circuit diagram and input and output waveforms of an inverting comparator with a 0V reference voltage.

Zero Crossing Detector Using UA 741 op-amp IC



## Zero - Crossing Detector Using 741 IC Waveforms



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❖ **MCQ** The zero crossing detector circuit plays a crucial role in conversion of input sine wave into a perfect ----- at its output.

Options are : A) Triangular wave    B) Square wave    C) Saw-tooth wave    D) Pulse wave .

## Schmitt trigger or squaring circuit.

Figure shows an Inverting comparator with positive feedback.

This circuit converts an irregular-shaped waveform to a square wave.

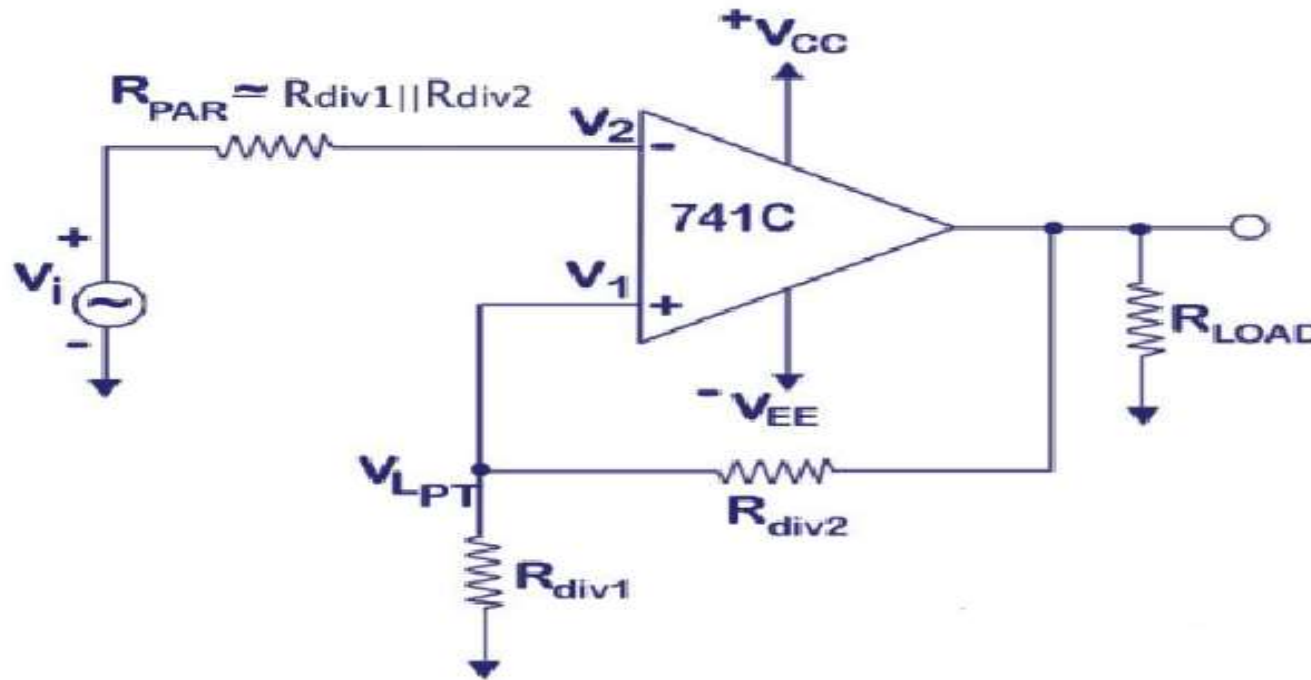
The circuit is known as **Schmitt trigger or squaring circuit**.

The input voltage  $V_{in}$  triggers the output  $V_o$  every time it exceeds certain voltage levels called the **upper threshold voltage  $V_{ut}$  & lower threshold voltage  $V_{lt}$** .

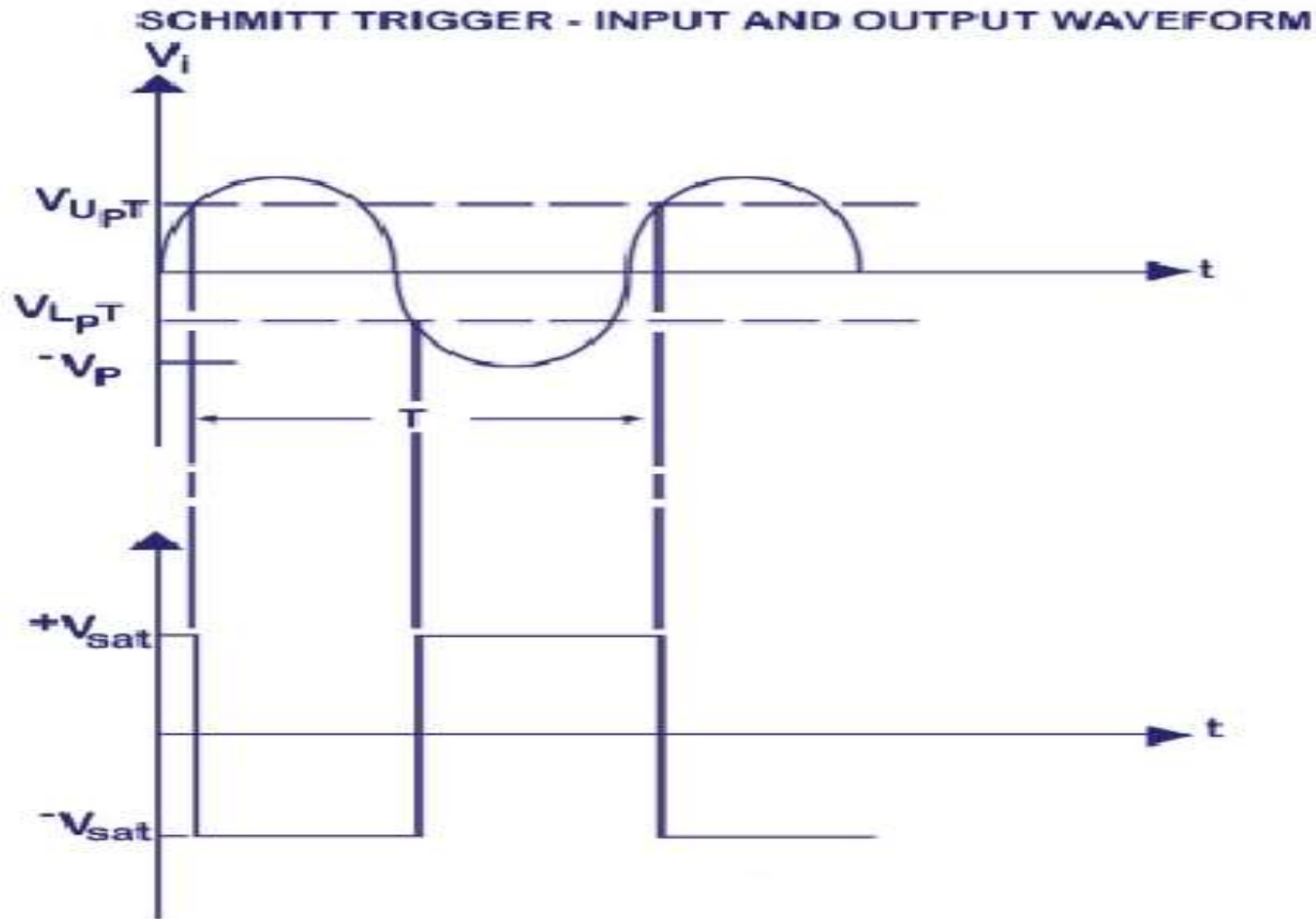
Where,  $V_{ut} = [R_1 / (R_1 + R_2)] * (+V_{sat}) =$  **upper threshold voltage  $V_{ut}$**

$V_{lt} = [R_1 / (R_1 + R_2)] * (-V_{sat}) =$  **lower threshold voltage  $V_{lt}$** .

SCHMITT TRIGGER USING OP - AMP 741IC



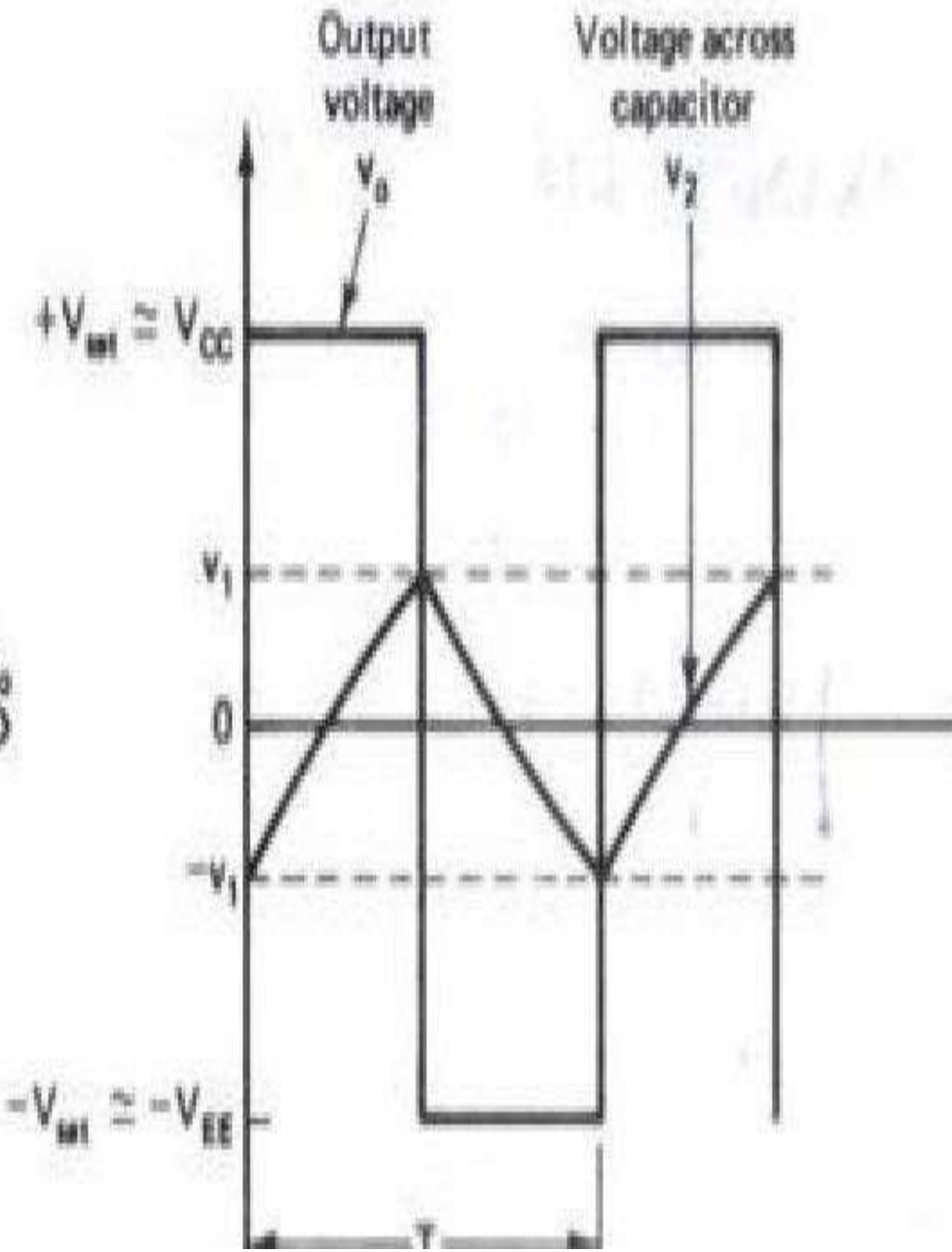
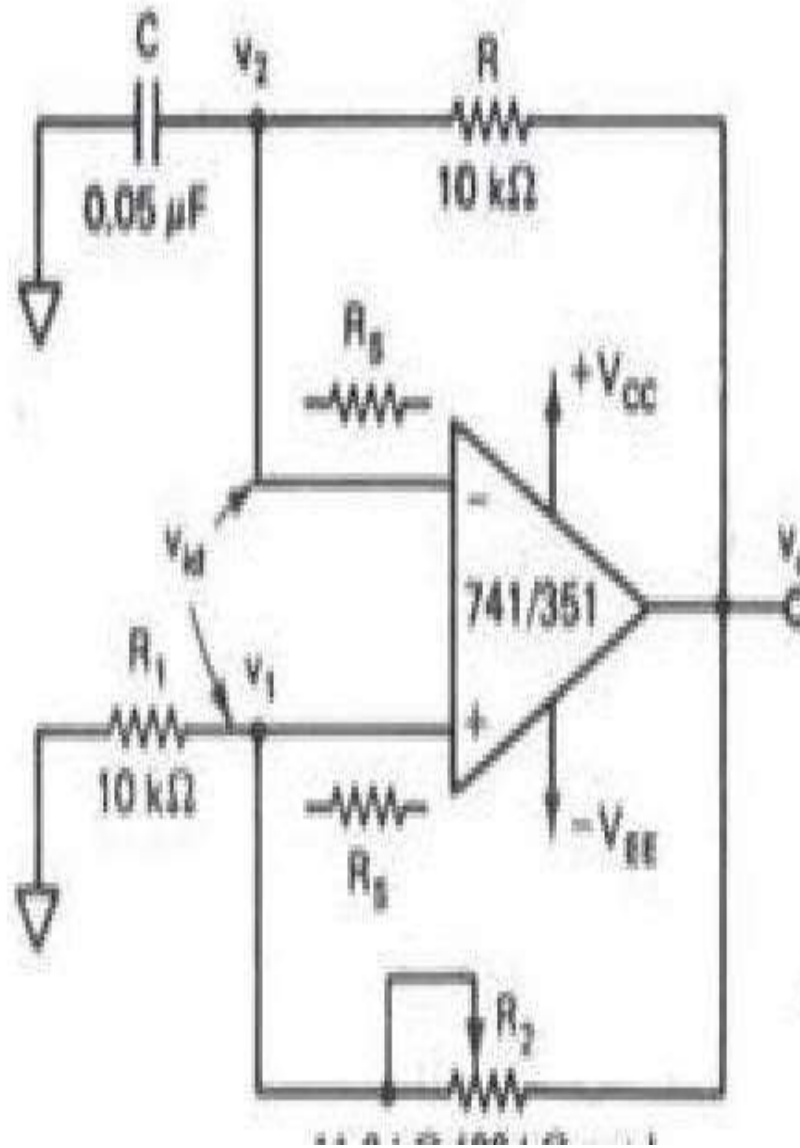




- ❖ MCQ : Which circuit converts irregularly shaped waveform to regular shaped waveforms?  
Options are : A) Schmitt trigger B) Voltage limiter C) Comparator D) Subtractor
- ❖ MCQ : When a large sine wave drives a Schmitt trigger, the output is a Rectangular wave  
Options are : A) Rectangular wave B) Triangular wave B) Rectified sine wave C) Series of ramps

# Square Wave Generator using OP-AMP 741 IC :

This square wave generator is also called a **free-running** or a **stable multi vibrator**. The output of the op-amp in this circuit will be in positive or negative saturation, depending on whether the differential voltage **vid** is negative Or positive, respectively



- **Working of the Circuit :**
- Assume that the voltage across capacitor C is zero volts at the instant the dc supply voltages +VCC and -VEE are applied.
- This means that the voltage at the **Inverting terminal** is zero initially. At the same instant, however, the voltage V1 at then **Non-inverting terminal** is a very small finite value that is a function of the output offset voltage V<sub>OOT</sub> and the values of R1 and R2 resistors.
- Thus the differential input voltage V<sub>id</sub> is equal to the voltage V1 at the Non-inverting terminal. Although very small, voltage V1 will start to drive the op-amp into saturation.
- For example, suppose that the output offset voltage V<sub>OOT</sub> is positive and that, therefore, voltage V1 is also positive.
- Since initially the capacitor C acts as a short circuit, the gain of the op-amp is very large (A); hence V1 drives the output of the op-amp to its positive saturation **+V<sub>sat</sub>**.
- With the output voltage of the op-amp at +V<sub>sat</sub>, the capacitor C starts charging toward +V<sub>sat</sub> through resistor R.
- However, as soon as the voltage V<sub>2</sub> across capacitor C is slightly more positive than V1, the output of the op-amp is forced to switch to a negative saturation, -V<sub>sat</sub>. With the op-amp's output voltage at negative saturation, -V<sub>sat</sub>, the voltage v1 across R1 is also negative, since

## Working of circuit:

- For generation of square wave the OPAMP is forced to operate in its saturation region.
- The voltage available at **Non-inverting input of OP-AMP** is obtained by potential-divider action.

$$V_1 = (R_1 / (R_1 + R_2)) V_o \text{ ----- } V_o = +V_{sat} \text{ initially}$$

Now,  $V_1 = \beta * V_o$ , where  $\beta = R_1 / (R_1 + R_2) = \text{feedback factor}$

- The o/p  $V_o$ , is also feedback to Inverting I/p terminal where I/p voltage obtained is given by

$$V_c = V_2 = (1/RC) \int V_o dt.$$

- Whenever  $V_c > V_1 > \beta V_o$  Switching takes place and square wave output is available from comparator developed by op-amp.
- Theoretically:  $f_o = 1 / 2RC$  ----- (Assuming,  $R_2 = 2R_1$ )
- Practically:  $f_o = 1 / T$

• **MCQ : Q1)** How are the square wave output generated in op-amp?

**Options are:** A) Op-amp is forced to operate in the positive saturation region

B) Op-amp is forced to operate in the negative saturation region

C) Op-amp is forced to operate in the positive and negative saturation region

D) None of the mentioned

**Q2)** An Astable multivibrator using op-amp has  $f_o = 1\text{kHz}$ . Assume the resistor value to be  $10\text{k}\Omega$ , find the capacitor value. Use  $R_2 = 1.16 R_1$

Options are: A)  $3.9 \mu\text{F}$       B)  $0.3 \mu\text{F}$       C)  $2\mu\text{F}$       D)  $0.05 \mu\text{F}$

**Q3)** To design a square wave generator using OP-AMP 741 IC for  $f = 700\text{Hz}$ .

## Derivation for frequency of Square wave

The frequency is determined by the time it takes the capacitor to charge from  $-\beta V_{sat}$  to  $+\beta V_{sat}$  and vice versa. The voltage across the capacitor as a function of time is given by,

$$v_c(t) = V_f + (V_i - V_f)e^{-t/RC} \quad (5.4)$$

where, the final value,  $V_f = +V_{sat}$   
and the initial value,  $V_i = -\beta V_{sat}$

Therefore,

$$v_c(t) = V_{sat} + (-\beta V_{sat} - V_{sat})e^{-t/RC} \quad (5.5)$$

or

$$v_c(t) = V_{sat} - V_{sat}(1 + \beta)e^{-t/RC}$$

At  $t = T_1$ , voltage across the capacitor reaches  $\beta V_{sat}$  and switching takes place. Therefore,

$$v_c(T_1) = \beta V_{sat} = V_{sat} - V_{sat}(1 + \beta)e^{-T_1/RC} \quad (5.6)$$

After algebraic manipulation, we get,

$$T_1 = RC \ln \frac{1 + \beta}{1 - \beta} \quad (5.7)$$

This gives only one half of the period.

Total time period

$$T = 2T_1 = 2RC \ln \frac{1 + \beta}{1 - \beta} \quad (5.8)$$

and the output wave form is symmetrical.

If  $R_1 = R_2$ , then  $\beta = 0.5$ , and  $T = 2RC \ln 3$ . And for  $R_1 = 1.16R_2$ , it can be seen that

$$T = 2RC$$

or

$$f_0 = \frac{1}{2RC}$$

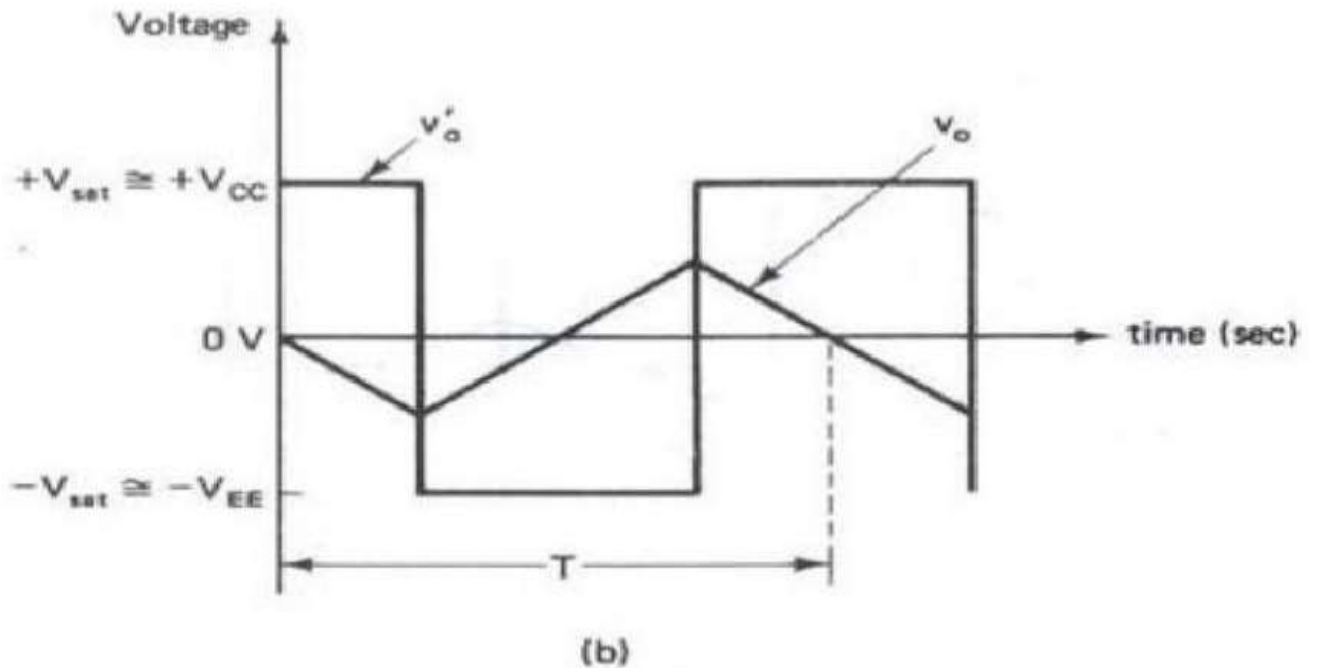
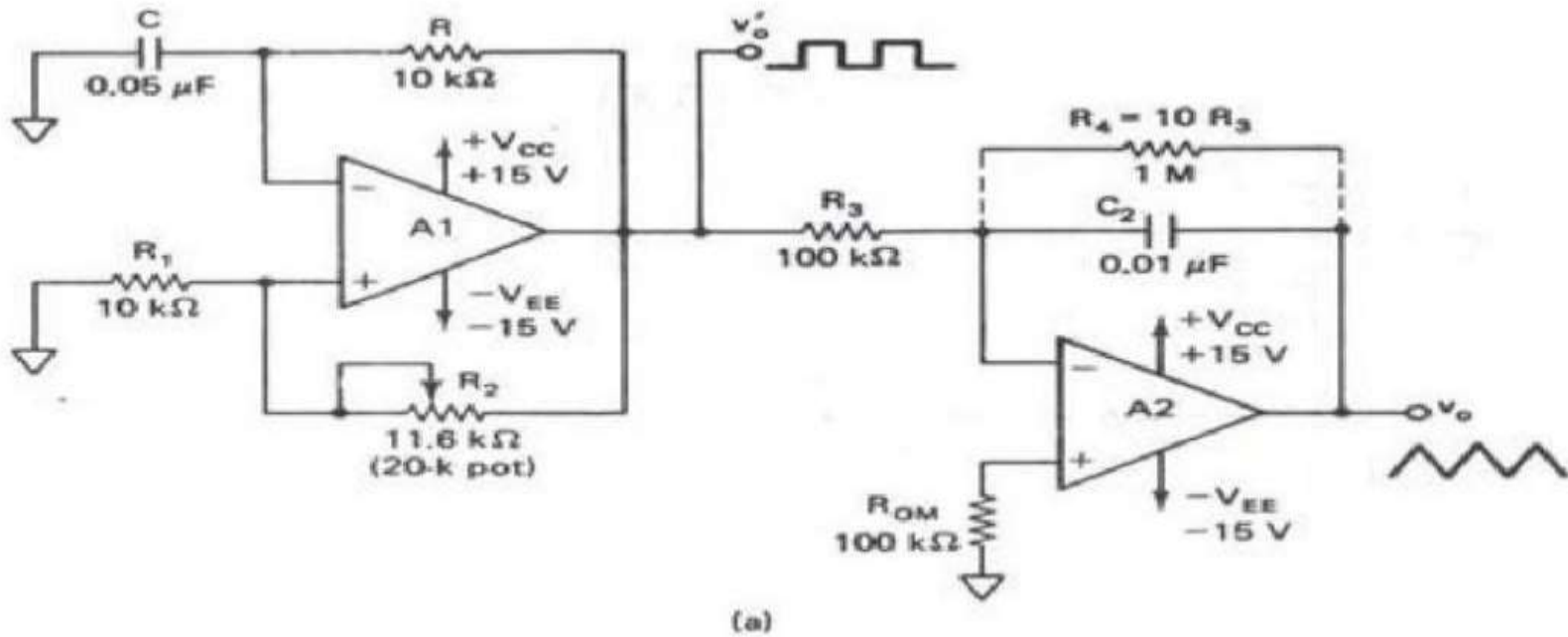
The output swings from  $+V_{sat}$  to  $-V_{sat}$ , so,

$$v_o \text{ peak-to-peak} = 2V_{sat} \quad (5.9)$$

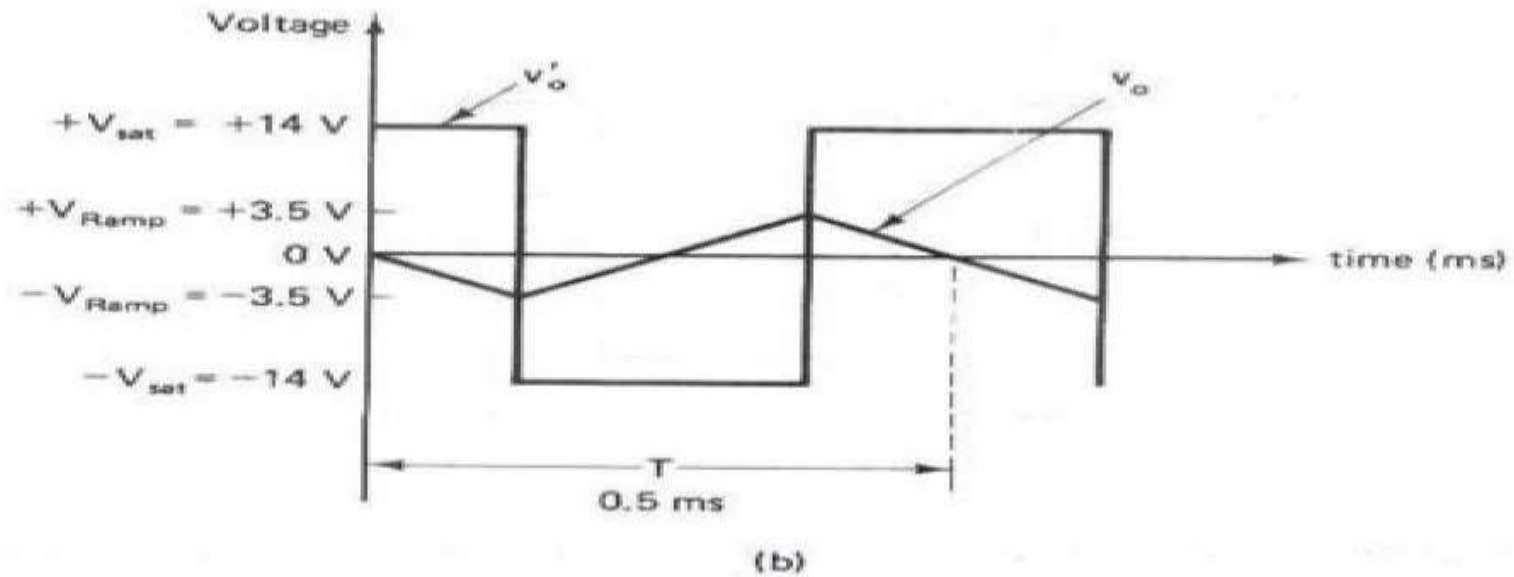
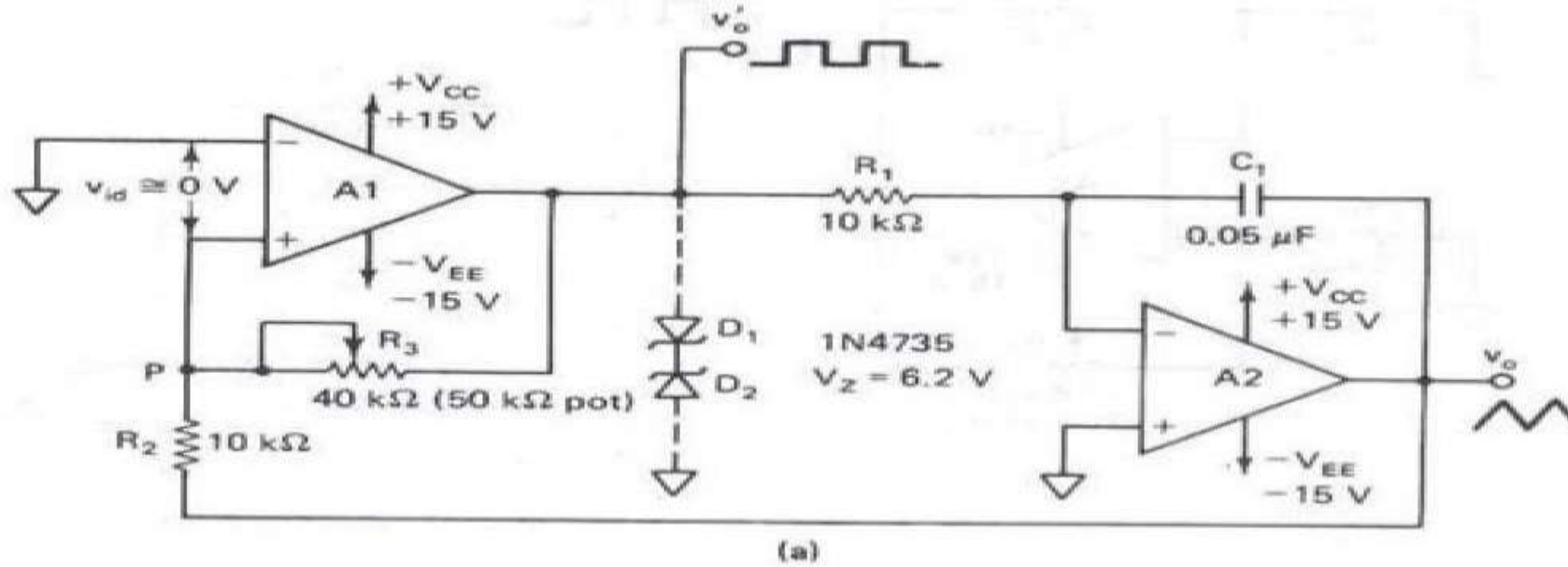
## Triangular Wave Generator using Square wave generator and Integrator

Recall that the output waveform of the integrator is triangular if its input is a square wave. This means that a triangular wave generator can be formed by simply connecting an integrator to the square wave generator. The resultant circuit is shown in Figure. This circuit requires a dual op-amp, two capacitors, and at least five resistors.

The frequencies of the square wave and triangular wave are the same. For fixed  $R_1$ ,  $R_2$ , and  $C$  values, the frequency of the square wave as well as the triangular wave depends on the resistance  $R$ .



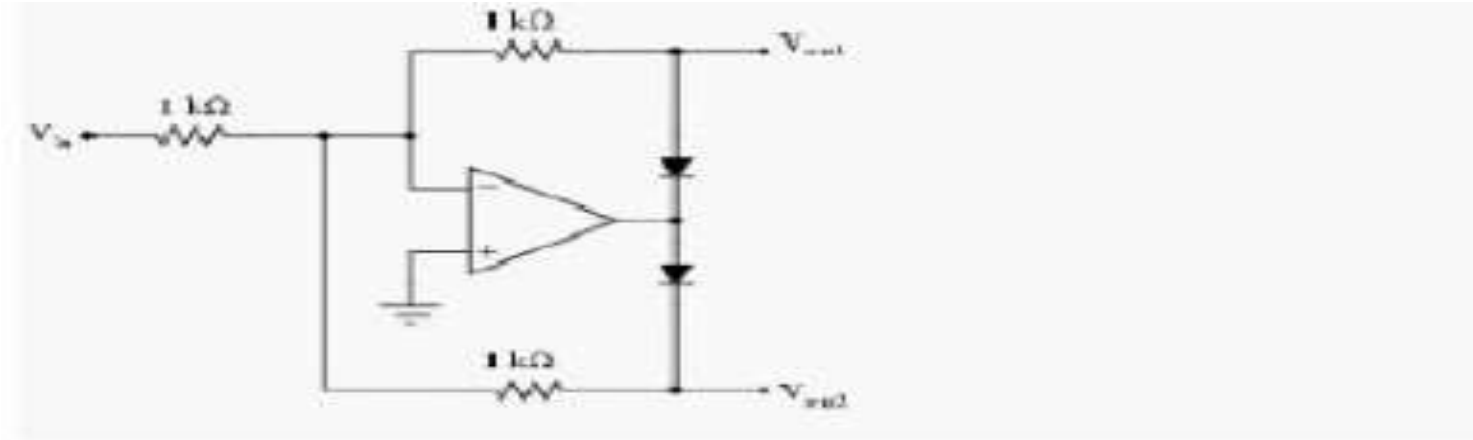
# Triangular Wave Generator using Comparator and Integrator





## Precision Rectifiers types : 1) Full wave Rectifier

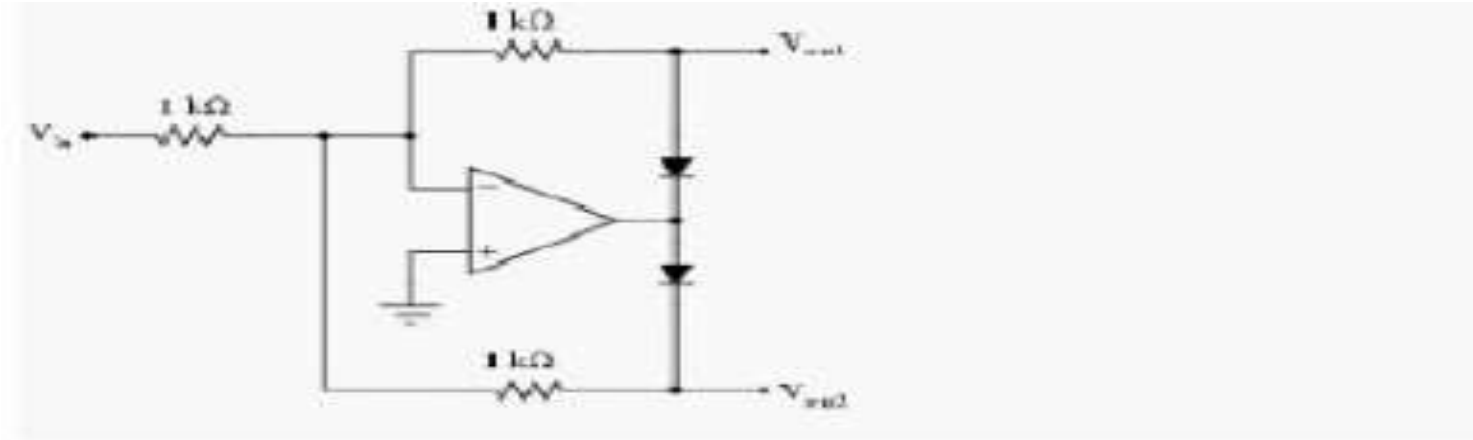
**MCQ.** Select the following option for following circuit, if Diodes are ideal and  $V_{in} = +V_p$



- Options are:**
- A)  $V_{out1} = -2V_p$ ,  $V_{out2} = 0\text{ Vp}$
  - B)  $V_{out1} = 2V_p$ ,  $V_{out2} = -2\text{ Vp}$
  - C)  $V_{out1} = 0\text{ Vp}$ ,  $V_{out2} = -2\text{ Vp}$
  - D)  $V_{out1} = 0\text{ Vp}$ ,  $V_{out2} = 2\text{ Vp}$

## Precision Rectifiers types : 1) Full wave Rectifier

**MCQ.** Select the following option for following circuit, if Diodes are ideal and  $V_{in} = + V_p$

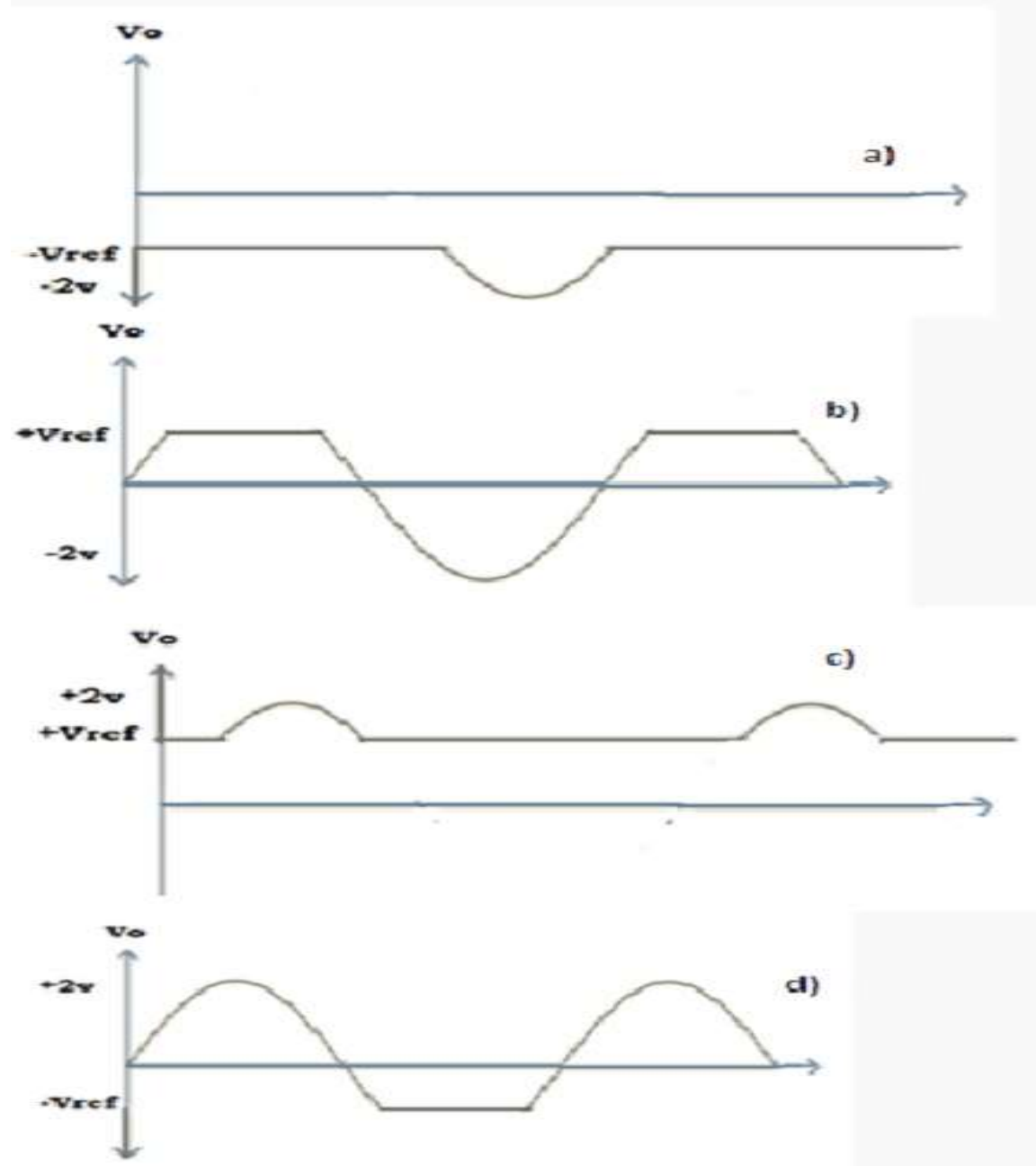
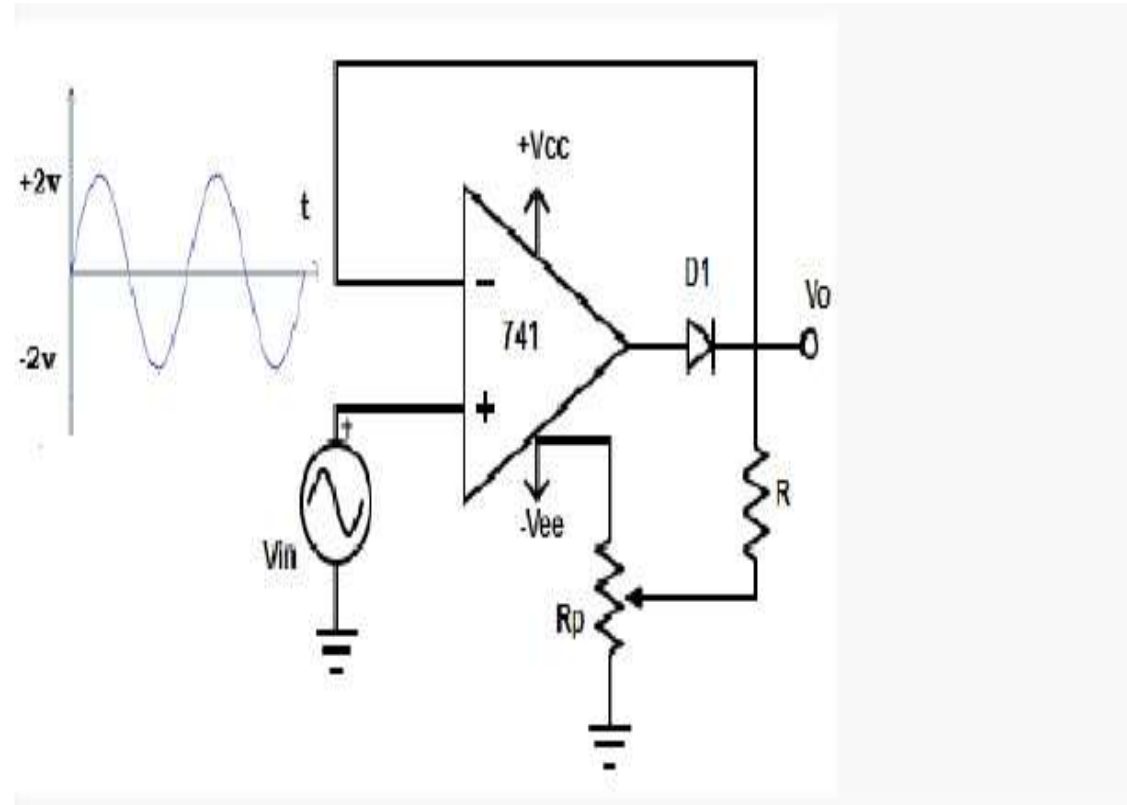


- Options are:**
- A)  $V_{out1} = -2V_p$ ,  $V_{out2} = 0\text{ Vp}$
  - B)  $V_{out1} = 2V_p$ ,  $V_{out2} = -2\text{ Vp}$
  - C)  $V_{out1} = 0\text{ Vp}$ ,  $V_{out2} = -2\text{ Vp}$
  - D)  $V_{out1} = 0\text{ Vp}$ ,  $V_{out2} = 2\text{ Vp}$

# Precision Rectifier:

## 2) Half wave rectifier

**MCQ:** For the circuit shown below Find the output Waveforms.



**Ans : d**

## Sample and Hold Circuits :

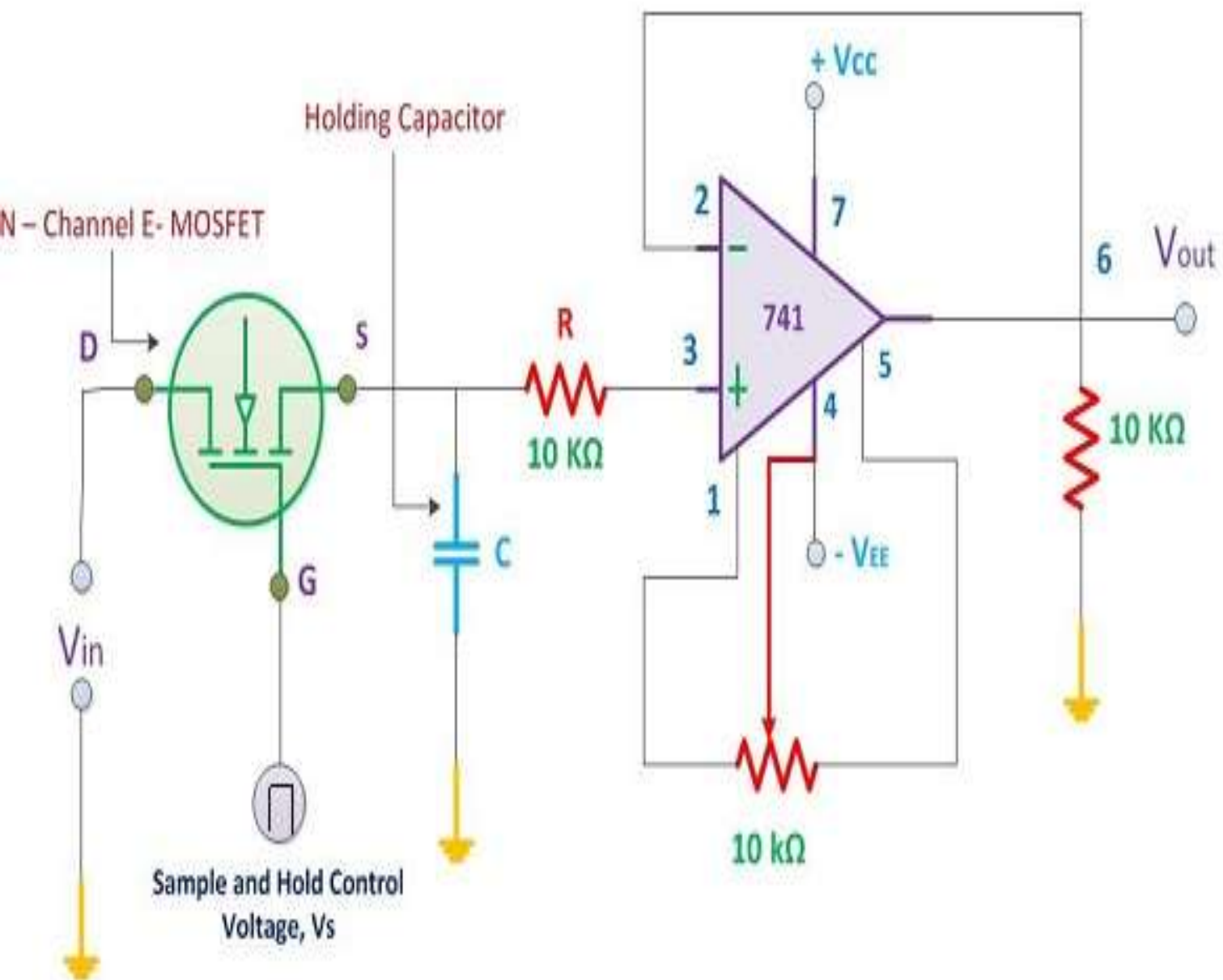
The sample and hold circuit, as its name implies samples an i/p signal and holds on to its last sampled value until the i/p is sampled again.

Below fig shows a sample and hold circuit using an op-amp with an E-MOSFET. In this circuit the E-MOSFET works as a switch that is controlled by the sample and control voltage  $V_s$ , and the capacitor  $C$  serves as a storage element.

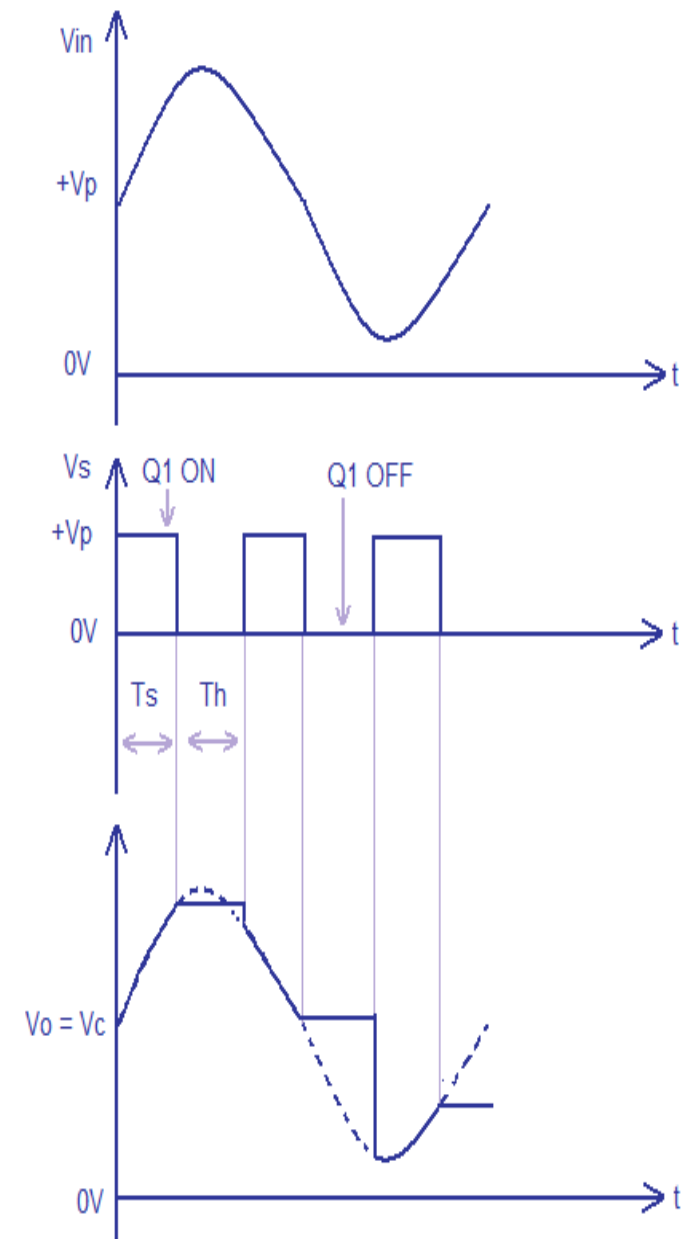
The analog signal  $V_{in}$  to be sampled is applied to the drain, and sample and hold control voltage  $V_s$  is applied to the gate of the E-MOSFET. During the positive portion of the  $V_s$ , the E-MOSFET conducts and acts as a closed switch. This allows i/p voltage to charge capacitor  $C$ . In other words input voltage appears across  $C$  and in turn at the o/p as shown in above fig.

On the other hand, when  $V_s$  is zero, the E-MOSFET is off and acts as open switch. The only discharge path for  $C$  is, through the op-amp. However the i/p resistance of the op-amp voltage follower is also very high; hence the voltage across  $C$  is retained.

The time periods  $T_s$  of the sample-and-hold control voltage  $V_s$  during which the voltage across the capacitor is equal to the i/p voltage are called sample periods. The time periods  $T_H$  of  $V_s$  during which the voltage across the capacitor is constant are called hold periods. The o/p of the op-amp is usually processed/ observed during hold periods. To obtain the close approximation of the i/p waveform, the frequency of the sample-and-hold control voltage must be significantly higher than that of the i/p.

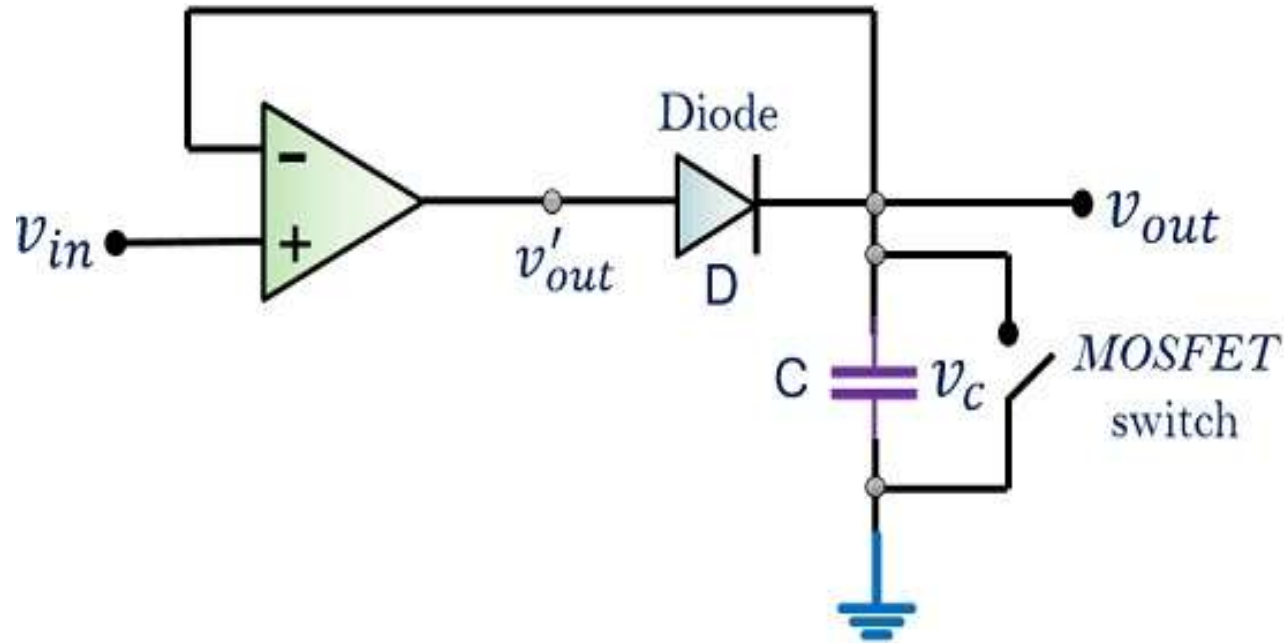


Practical Circuit of Sample and Hold Circuit



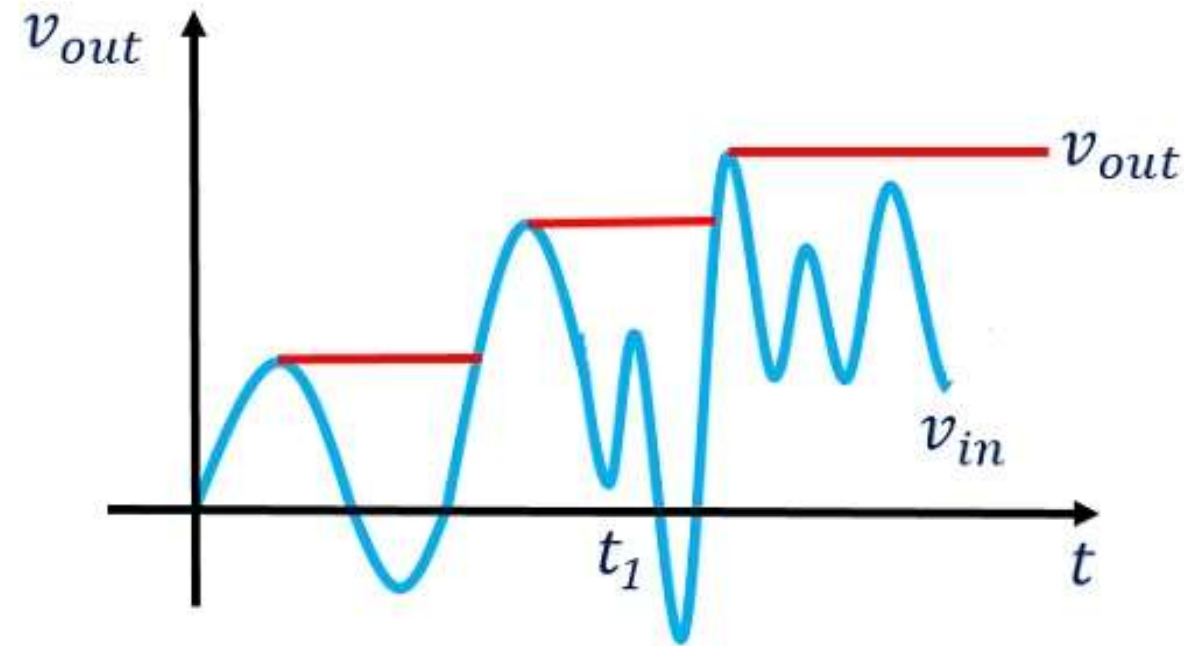
input and output waveforms

## Peak detector circuit:



Peak detector circuit

Electronics Coach



Output waveform for an applied input waveform

Electronics Coach

**MCQ: Q1) Among which of the following factors do the operation of sample and hold mode depend?**

**A) Input    B) Output    C) Position of switch    D) TIME**

**Q2) In a peak detector circuit, which component holds the peak value till a higher peak value is detected ?**

**Options are : A) Diode    B) Inductor    C) Capacitor    D) MOSFET switch**

Q3. The most commonly used amplifier in sample and hold circuit is

- A) a unity gain inverting amplifier
- B) a unity gain non-inverting amplifier
- C) an inverting amplifier with a gain of 10
- D) an inverting amplifier with a gain of 100